

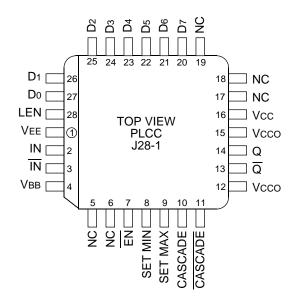


# PROGRAMMABLE DELAY CHIP

### FEATURES

- Up to 2ns delay range
- Extended 100E VEE range of -4.2V to -5.5V
- ≈20ps/digital step resolution
- >1GHz bandwidth
- On-chip cascade circuitry
- 75KkΩ input pulldown resistor
- Fully compatible with Motorola MC10E/100E195
- Available in 28-pin PLCC package

### PIN CONFIGURATION



### DESCRIPTION

The SY10/100E195 are programmable delay chips (PDCs) designed primarily for clock de-skewing and timing adjustment. They provide variable delay of a differential ECL input transition.

The delay section consists of a chain of gates organized as shown in the logic diagram. The first two delay elements feature gates that have been modified to have delays 1.25 and 1.5 times the basic gate delay of approximately 80ps. These two elements provide the E195 with a digitally-selectable resolution of approximately 20ps. The required device delay is selected by the seven address inputs D[0:6], which are latched on-chip by a high signal on the latch enable (LEN) control. If the LEN signal is either LOW or left floating, then the latch is transparent.

Because the delay programmability of the E195 is achieved by purely differential ECL gate delays, the device will operate at frequencies of >1GHz, while maintaining over 600mV of output swing.

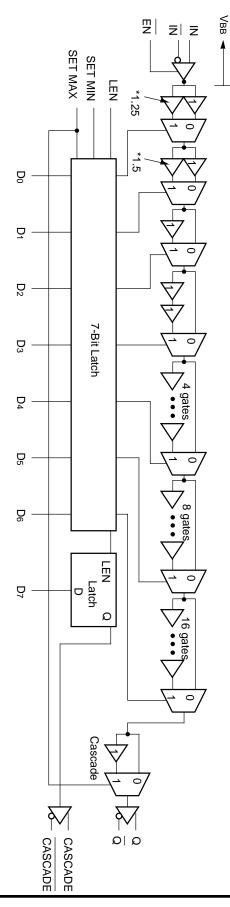
The E195 thus offers very fine resolution, at very high frequencies, selectable entirely from a digital input, allowing for very accurate system clock timing.

An eighth latched input, D7, is provided for cascading multiple PDCs for increased programmable range. The cascade logic allows full control of multiple PDCs, at the expense of only a single added line to the data bus for each additional PDC, without the need for any external gating.

### **PIN NAMES**

Pin	Function				
IN/ĪN	Signal Input				
ĒN	Input Enable				
D[0:7]	Mux Select Inputs				
$Q/\overline{Q}$	Signal Output				
LEN	Latch Enable				
SET MIN	Minimum Delay Set				
SET MAX	Maximum Delay Set				
CASCADE	Cascade Signal				

# **BLOCK DIAGRAM**



\*Delays are 25% or 50% longer than standard (standard = 80ps).

# DC ELECTRICAL CHARACTERISTICS

VEE = VEE (Min.) to VEE (Max.); VCC = GND

		TA = 0°C		TA = +25°C			TA = +85°C					
Symbol	Parameter	Min.	Тур.	Max.	Min.	Тур.	Max.	Min.	Тур.	Max.	Unit	Condition
Ін	Input HIGH Current	_	_	150	_	—	150	—		150	μA	_
IEE	Power Supply Current										mA	_
	10E	—	130	156	—	130	156	—	130	156		
	100E	—	130	156	—	130	156	—	150	179		

## **AC ELECTRICAL CHARACTERISTICS**

VEE = VEE (Min.) to VEE (Max.); VCC = GND

		TA = 0°C		TA = +25°C			TA = +85°C					
Symbol	Parameter	Min.	Тур.	Max.	Min.	Тур.	Max.	Min.	Тур.	Max.	Unit	Condition
tplh tphl	Propagation Delay to Output IN to Q; Tap = 0 IN to Q; Tap = 127 $\overline{EN}$ to Q; Tap = 0 D7 to CASCADE	1210 3320 1250 300	1360 3570 1450 450	1510 3820 1650 700	1240 3380 1275 300	1390 3630 1475 450	1540 3880 1675 700	1440 3920 1350 300	1590 4270 1650 450	1765 4720 1950 700	ps	_
trange	Programmable Range tPD (max.) – tPD (min.)	2000	2175		2050	2240		2375	2580		ps	—
Δt	Step Delay Do High D1 High D2 High D3 High D4 High D5 High D6 High	 55 115 250 505 1000	17 34 68 136 272 544 1088	— 105 180 325 620 1190	 55 115 250 515 1030	17.5 35 70 140 280 560 1120	 105 180 325 620 1220	 65 140 305 620 1240	21 42 84 168 336 672 1344	— 120 205 380 740 1450	ps	6
Lin	Linearity	D1	D0		D1	D0	_	D1	D0	_		7
tskew	Duty Cycle Skew, tPHL-tPLH		±30	_		±30	_	_	±30		ps	1
ts	Set-up Time D to LEN D to IN EN to IN	200 800 200	0 		200 800 200	0		200 800 200	0		ps	2 3
tн	Hold Time LEN to D IN to EN	500 0	250 —		500 0	250 —		500 0	250 —		ps	4
tR	Release Time EN to IN SET MAX to LEN SET MIN to LEN	300 800 800			300 800 800			300 800 800			ps	5
tjit	Jitter	—	<5	—	—	<5	_	_	<5	—	ps	8
tr tf	Rise/Fall Times 20–80% (Q) 20–80% (CASCADE)	125 300	225 450	325 650	125 300	225 450	325 650	125 300	225 450	325 650	ps	—

#### NOTES:

2. Duty cycle skew guaranteed only for differential operation measured from the cross point of the input to the cross point of the output.

3. This set-up time defines the amount of time prior to the input signal the delay tap of the device must be set.

- 4. This set-up time is the minimum time that EN must be asserted prior to the next transition of IN/IN to prevent an output response greater than ±75mV to that IN/IN transition.
- 5. This hold time is the minimum time that EN must remain asserted after a negative going IN or positive going IN to prevent an output response greater than ±75mV to that IN/IN transition.
- 6. This release time is the minimum time that EN must be deasserted prior to the next IN/IN transition to ensure an output response that meets the specified IN to Q propagation delay and transition times.
- 7. Specification limits represent the amount of delay added with the assertion of each individual delay control pin. The various combinations of asserted delay control inputs will typically realize Do resolution steps across the specified programmable range.
- 8. The linearity specification guarantees to which delay control input the programmable steps will be monotonic (i.e. increasing delay steps for increasing binary counts on the control inputs Dn). Typically, the device will be monotonic to the Do input, however, under worst case conditions and process variation, delays could decrease slightly with increasing binary counts when the Do input is the LSB. With the D1 input as the LSB, the device is guaranteed to be monotonic over all specified environmental conditions and process variation.

9. The jitter of the device is less than what can be measured without resorting to very tedious and specialized measurement techniques.

#### **APPLICATIONS INFORMATION**

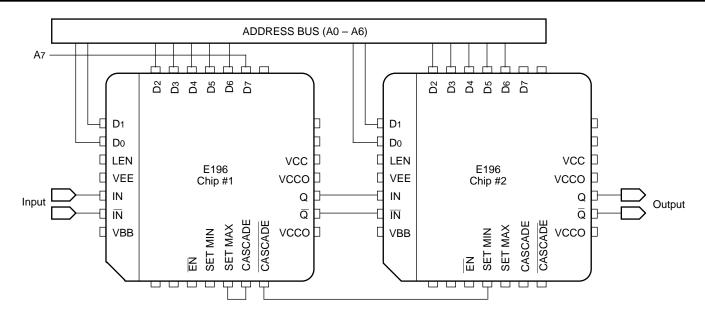


Figure 1. Cascading Interconnect Architecture

#### **Cascading Multiple E195s**

To increase the programmable range of the E195, internal cascade circuitry has been included. This circuitry allows for the cascading of multiple E195s without the need for any external gating. Furthermore, this capability requires only one more address line per added E195. Obviously, cascading multiple PDCs will result in a larger programmable range; however, this increase is at the expense of a longer minimum delay.

Figure 1 illustrates the interconnect scheme for cascading two E195s. As can be seen, this scheme can easily be expanded for larger E195 chains. The D7 input of the E195 is the cascade control pin. With the interconnect scheme of Figure 1, when D7 is asserted, it signals the need for a larger programmable range than is achievable with a single device.

An expansion of the latch section of the block diagram is pictured below. Use of this diagram will simplify the explanation of how the cascade circuitry works. When D7 of chip #1 above is low, the cascade output will also be low, while the cascade bar output will be a logical high. In this condition, the SET MIN pin of chip #2 will be asserted and, thus, all of the latches of chip #2 will be reset and the device will be set at its minimum delay. Since the RESET and SET inputs of the latches are overriding, any changes on the A0–A6 address bus will not affect the operation of chip #2. Chip #1, on the other hand, will have both SET MIN and SET MAX de-asserted so that its delay will be controlled entirely by the address bus  $A_0$ -A6. If the delay needed is greater than can be achieved with 31.75 gate delays (1111111 on the A0-A6 address bus), D7 will be asserted to signal the need to cascade the delay to the next E195 device. When D7 is asserted, the SET MIN pin of chip #2 will be de-asserted and the delay will be controlled by the A0-A6 address bus. Chip #1, on the other hand, will have its SET MAX pin asserted, resulting in the device delay to be independent of the A0-A6 address bus.

When the SET MAX pin of chip #1 is asserted, the Do and D1 latches will be reset while the rest of the latches will be set. In addition, to maintain monotonicity, an additional gate delay is selected in the cascade circuitry. As a result, when D7 of chip #1 is asserted, the delay increases from 31.75 gates to 32 gates. A 32-gate delay is the maximum delay setting for the E195.

To expand this cascading scheme to more devices, one simply needs to connect the D7 input and CASCADE outputs of the current most significant E195 to the new most significant E195 in the same manner as pictured in Figure 1. The only addition to the logic is the increase of one line to the address bus for cascade control of the second PDC. SET MIN SET MAX LEN  $\mathsf{D}_0$ eset Bit 0 λ Q LEN Ā Bit 1 J õ LEN  $D_2$ Bit 2 Q LEZ D3 Set Bit 3 Reset Q LEN **D**<sub>4</sub> Set Bit 4 Reset Q<sub>4</sub> E D5 Se Bit 5 Reset Q5 LEN  $\mathsf{D}_6$ Set Bit 6 Re Q LEN D7 Bit 7 Q CASCADE CASCADE

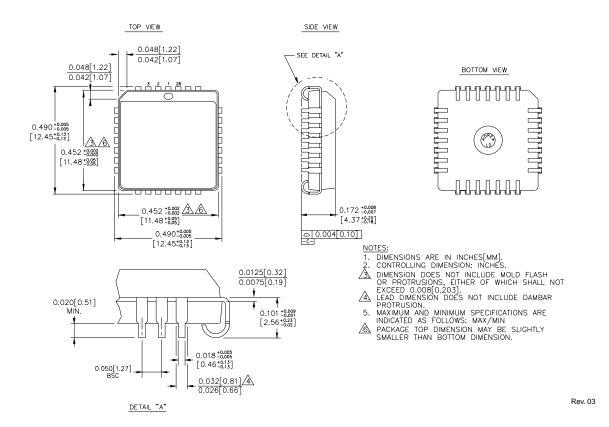
To Select Multiplexers

Figure 2. Expansion of the Latch Section of the E195 Block Diagram

# PRODUCT ORDERING CODE

Ordering Code	Package Type	Operating Range
SY10E195JC	J28-1	Commercial
SY10E195JCTR	J28-1	Commercial
SY100E195JC	J28-1	Commercial
SY100E195JCTR	J28-1	Commercial

## 28 LEAD PLCC (J28-1)



#### MICREL-SYNERGY 3250 SCOTT BOULEVARD SANTA CLARA CA 95054 USA

TEL + 1 (408) 980-9191 FAX + 1 (408) 914-7878 WEB http://www.micrel.com

This information is believed to be accurate and reliable, however no responsibility is assumed by Micrel for its use nor for any infringement of patents or other rights of third parties resulting from its use. No license is granted by implication or otherwise under any patent or patent right of Micrel Inc. © 2000 Micrel Incorporated